

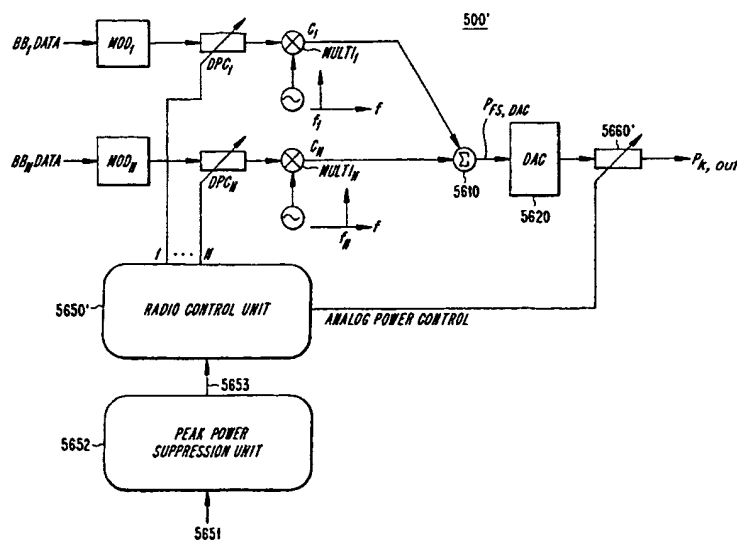
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- (71) Applicant: **TELEFONAKTIEBOLAGET LM ERICSSON (publ)** [SE/SE]; S-126 25 Stockholm (SE).
- (72) Inventors: **HEDBERG, Bo**; Kastrupgatan 10, S-164 41 Kista (SE). **FRODIGH, Magnus**; Risvägen 1F, S-192 73 Sollentuna (SE). **JOHANSSON, Christer**; Framnäsbacken 18, S-171 66 Solna (SE). **ERIKSSON, Patrik**; Thoresta Gård, S-197 93 Bro (SE).
- (74) Agents: **MILDH, Christer et al.**; Ericsson Radio Systems AB, Ericsson Research/Patent Support Unit, S-164 80 Stockholm (SE).
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(54) Title: TRANSMIT POWER CONTROL FOR MCPA-EQUIPPED BASE STATIONS



(57) Abstract: A transmit power control system and method is provided which utilizes an efficient allocation of average output power of a multi-carrier power amplifier (MCPA), the average output power being a weighted average of the actual power over multiple time slots. Periodically, using the efficient allocation system and method, there will be short periods where the desired average output power will exceed the maximum tolerable power of the MCPA, P_{MCPA} . These short periods in which the desired output power will exceed P_{MCPA} are handled by a system and method which produces a reduced margin between the total power per time slot and P_{MCPA} . This reduced margin allows the MCPA to serve a larger number of users per time slot than conventional MCPAs. Alternatively, the MCPA can serve the same number of users as a conventional MCPA, but with higher demands on output power for each user-and/or-more-efficient use of the MCPA's resources.

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TRANSMIT POWER CONTROL FOR MCPA-EQUIPPED BASE STATIONS

CROSS-REFERENCE TO RELATED APPLICATIONS

This application is related to co-pending U.S. Patent Application No.
5 09/349,717, filed on July 8, 1999, entitled "Power Control Scheme for
Maximizing Carrier Signal-to-Noise Ratio in Multicarrier Transmitters," which is
incorporated herein by reference in its entirety.

BACKGROUND

The present invention relates generally to cellular communication systems.
10 More precisely, the present invention relates to a transmit power supervision
method for a base station equipped with a multi-carrier power amplifier (MCPA).

In conventional cellular systems, a base station is allocated a predetermined
number of frequency channels for communication with mobile stations. In the
base station a separate transmitter is employed for each frequency channel.
15 However, the use of separate transmitters for each frequency channel results in a
duplication of parts and an increase in cost due to the additional hardware
required. Thereafter, it was realized that the hardware cost per channel could be
reduced by using multicarrier transmitters in place of the plurality of single carrier
transmitters to transmit a plurality of frequency channels. Since multicarrier
20 transmitters transmit over a broad range of frequencies, they are also sometimes
referred to in the art as wideband transmitters. However, for ease of discussion,
the transmitters will be referred to herein as multicarrier transmitters.

FIG. 1 illustrates a conventional multicarrier transmitter 100 which may be
used to transmit multiple frequency channels from a base station in a
25 radiocommunication system. The conventional multicarrier transmitter 100
operates as follows. A number N of baseband frequency data signals BB_1, \dots, BB_N
are modulated by modulators Mod_1, \dots, Mod_N , respectively, where the bits
associated with each data signal are symbol encoded for transmission, i.e., the

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modulator generates the corresponding baseband waveform. Each of the modulated data signals is forwarded to a corresponding digital power control module DPC_1, \dots, DPC_N , where each DPC adjusts the signal power level of the corresponding modulated data signal based on the commands provided by the

5 Radio Control Unit 150. More specifically, the power level of each modulated data signal is adjusted such that the absolute power level of each carrier $P_{k,out}$ at the transmitter is equal to the amount of power required for the carrier to reach a particular mobile station which is to receive the carrier, where k varies from 1 to N and identifies the corresponding baseband frequency data signals BB_1, \dots, BB_N .

10 The modulated data signals are then forwarded from the digital power control modules DPC_1, \dots, DPC_N to multipliers $Mult_1, \dots, Mult_N$, respectively, where each modulated data signal is upconverted to a corresponding carrier frequency. The upconverted signals are then summed by adder 110. The compound signal produced by adder 110 is then forwarded to the digital-to-analog converter (DAC)

15 120. The resulting compound analog signal is then passed from DAC 120 through an analog transmitter chain which includes analog amplifier 160, upconverter (not shown), and filters (not shown). Analog amplifier 160 then amplifies the compound signal by a fixed gain G_{ana} . For ease of discussion G_{ana} has been described as the gain of analog amplifier 160, however, one skilled in the art will

20 recognize that G_{ana} represents the total gain of the analog section of the transmitter, including losses due to filters and upconverters. A more detailed discussion of multicarrier transmitters can be found in "Base-Station Technology Takes Software-Definable Approach" by Richard M. Lober, Wireless System Designs, February 1998, which is herein incorporated by reference.

25 Multicarrier transmitters are designed to handle a maximum number of simultaneous carriers N . In designing a multicarrier transmitter, care must be taken to ensure that the instantaneous in-phase power sum, P_{sum} , of the N carriers does not exceed the maximum tolerable power of the MCPA. P_{sum} can be calculated using equation (1) below, where P_n represents the power of a specific

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user, n , in a specific time slot on a specific carrier frequency, and N is the total number of carrier frequencies used by the base station. Normally, P_n is equal to the peak power within the specific time slot.

The instantaneous in-phase power sum of a single time slot for a system
 5 having, for example, a constant envelope (such as a GSM system), is given by:

$$P_{sum} = \left(\sum_{n=1}^N \sqrt{P_n} \right)^2 \quad (1)$$

For example, if the instantaneous sum of the N carrier frequencies exceeds the full scale range of the DAC, i.e., the value associated with the greatest digital code that can be converted into an analog value, the DAC will clip the analog signal. Clipping, i.e., preventing the analog signal from exceeding the amplitude
 10 corresponding to the full scale range of the DAC will have an adverse effect on the quality of the transmitted signal. However, one skilled in the art will recognize that in practical applications, a system might tolerate a power level which exceeds the DAC's full scale range by a small amount for short periods of time without suffering a decrease in system performance.

15 In a multicarrier transmitter with N carrier frequencies, the abovementioned "clipping" of the analog signal can be avoided by setting the full scale range of the DAC to $20 \cdot \log(N)$ dB above the maximum allowed peak power level of any individual carrier $1 \dots N$, since the full scale range set $20 \cdot \log(N)$ dB above the maximum power level of any individual carrier represents the greatest
 20 power level attainable by the sum of the N carriers.

Designing MCPAs with a high output power is a difficult and expensive task. As the MCPA is designed to have a higher maximum output power, design costs become increasingly more expensive. For a base station operating using time division multiple access (TDMA), the maximum total output power of the

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base station limits the total output power of the frequency carriers at any time slot. TDMA, as one skilled in the art will appreciate, is a communication technique whereby different signals are assigned to different time slots on the same frequencies. One problem associated with MCPAs designed for a particular output power and operating in a TDMA environment is that a MCPA can only
5 serve a predetermined maximum number of users for the respective particular output power. If more than the predetermined maximum number of users were to be allocated to the MCPA, the MCPA would lose linearity resulting in a decrease in link quality.

10 FIG. 2 illustrates an exemplary time chart which may be associated with a base station. In FIG. 2, seven frequencies (1-7) in use by an exemplary base station are illustrated over eight time slots. The numbers in the time chart indicate the required output power, in watts, for a mobile unit which is operating at a particular frequency and assigned to a particular time slot. For example, at
15 frequency 1 and time slot 1, the mobile unit requires 4 watts (W). P_{sum} for each time slot is depicted below the time chart.

Assuming, for example, that the maximum power level that can be allocated for each mobile unit is set to 8W, then using Equation (1), the serving MCPA must be designed for at least a maximum output power of 392W. That is,
20 the MCPA must be designed to handle the worst-case scenario of each of the mobile units receiving at their maximum allocated power, 8W. As seen in FIG. 2, the in-phase power sum per time slot, P_{sum} , will usually be lower than the maximum output power of 392W. This difference illustrates how the MPCA will not be used efficiently, since it must be designed to handle a worst-case scenario
25 of all users being allocated the maximum output power for a given time slot.

Several techniques have been developed for extending the maximum capacity for which MCPAs have been dimensioned. Load sharing is one such technique. Conventional load sharing is basically a type of load balancing where a user is transferred from one cell which has reached its maximum capacity to

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another cell which can accommodate the user. This technique avoids overload situations. The following patents illustrate conventional load sharing techniques.

A method of balancing the load among cells which are operating at maximum capacity is described in U.S. Patent No. 4,670,899, by Brody et al.,
5 and entitled "Load Balancing for Cellular Radio Telephone System." In Brody et al., the loading of various cells is dynamically redistributed by selectively transferring ongoing calls to adjacent cells in accordance with traffic levels in order to reserve channels for handoffs and for new calls. A channel occupancy level for a cell is periodically determined by comparing the number of channels
10 utilized to the number of channels available within the cell. Calls are handed off before all the channels are utilized, thereby allowing at least one or more channels to be reserved for new or incoming calls.

According to the Brody et al. patent, if there is a mobile unit on the periphery of the cell which is also within the range of a neighboring cell, the
15 mobile unit will be transferred to the neighboring cell in order to make room for a new call or an ongoing call associated with a mobile unit which will be handed off to the cell. While Brody et al. provides traffic-based control for call handoffs from one cell to an adjacent cell, handoffs due to load balancing are handled differently from handoffs due to mobile units leaving the cell. This creates a very
20 complex system.

In U.S. Patent No. 5,625,868 to Jan et al., and entitled "Dynamic Traffic Load Distribution Method," the control over a call is transferred from a first satellite to a second satellite having a partially overlapping coverage area with the first satellite when the power consumption level in the first satellite exceeds a
25 certain predetermined level. This is accomplished by switching the channel off in the first satellite and on in the second satellite.

In commonly assigned U.S. Patent No. 5,241,685 to Bodin et al., and entitled "Load Sharing Control for a Mobile Cellular Radio System," the entirety of which is incorporated by reference herein, a load sharing method is set forth

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which is based upon the occupancy of the channels defined by the ratio between the number of occupied channels to the number of available channels.

In commonly assigned U.S. Patent Application No. 09/166,159 to Johansson et al., and entitled "Load Sharing for MCPA-Equipped Base Stations," the entirety of which is incorporated by reference herein, a method is set forth by which users are moved between time slots or between base stations in order to better utilize MCPA output power resources. The power of all transceivers served by a certain MCPA is compared to a threshold value. The threshold value is at least related to the capability of the MCPA in terms of output power. If at no times the sum of the output power exceeds the threshold value, it is assumed that the MCPA can handle all simultaneous transmissions. If, during at least one time slot it is found that the required output power exceeds the capability of the serving MCPA, then a reallocation or load sharing algorithm is invoked. The reallocation algorithm searches for time slots to which a reallocation could be performed within the same base station or number of transceivers served by the MCPA. If no time slots can handle users from the time slot in which the MCPA limit is exceeded, then a load sharing algorithm is activated and one starts to look for transmission resources in transceivers served by other MCPAs.

SUMMARY

The present invention distinguishes over the above techniques by providing a more efficient allocation of average output power for a MCPA, the average output power being a weighted average of the actual power over multiple time slots. Periodically, using the more efficient allocation system and method of the present invention, there will be short periods where the desired output power will exceed the maximum tolerable power of the MCPA, P_{MCPA} . These short periods in which the desired output power will exceed P_{MCPA} are handled by an exemplary embodiment of the present invention which produces a reduced margin between the average total power over several time slots and P_{MCPA} . This reduced margin

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allows the MCPA of the present invention to serve a larger number of users per time slot than conventional MCPAs. Alternatively, the MCPA of the present invention can serve the same number of users as a conventional MCPA, but with higher demands on output power for each user and/or more efficient use of the

5 MCPA's resources.

According to a first aspect of the present invention, a method for power control in a radio communication system is provided. The method includes determining a power requirement for an output signal; comparing the power requirement with a threshold power value; and modifying the output signal when

10 the power requirement is greater than the threshold power value such that the modified output signal has a power requirement which is less than or equal to the threshold power value.

According to a further aspect of the present invention, a method for power control in a radio communication system is provided. The method includes

15 allocating a respective power for at least one of a plurality of mobile terminals such that an instantaneous power of an output signal for said plurality of mobile terminals will exceed a threshold power value of a power amplifier; determining a power requirement for the output signal; comparing the power requirement for the output signal with the threshold power value; and modifying the output signal

20 when the power requirement is greater than the threshold power value such that the modified output signal has a modified power requirement which is less than or equal to the threshold power value.

According to a further aspect of the invention, a power control system for a radio communication system is provided. The power control system includes a

25 power supervision unit which determines a power requirement for a portion of an output signal and compares the power requirement with a threshold power value; and a signal control unit which modifies the output signal when the power requirement is greater than the threshold power value such that the modified

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output signal has a modified power requirement which is less than or equal to the threshold power value.

According to a further aspect of the invention, a power control system is provided. The power control system includes means for allocating a respective
5 power level for at least one of a plurality of mobile terminals such that an instantaneous power of an output signal for the plurality of mobile terminals will exceed a threshold power value of a power amplifier; means for determining a power requirement for the output signal; means for comparing the power
10 requirement of the output signal with the threshold power value; and means for modifying the output signal when the power requirement is greater than the threshold power value such that the modified output signal has a modified power requirement which is less than or equal to the threshold power value.

According to a further aspect of the invention, a power control system for a radio communication system is provided. The power control system includes a
15 power supervision unit which determines a power requirement for an output signal and compares the power requirement with a threshold power value. A signal control unit variably modifies the output signal in a digital domain and in an analog domain and a power suppression unit coupled to the signal control unit modifies the power requirement when the power requirement is greater than the
20 threshold power value such that the modified output signal has a modified power requirement which is less than or equal to the threshold power value.

According to a further aspect of the invention, a power control system for a radio communication system is provided. The power control system includes a digital power control module; a digital-to-analog converter; an analog power
25 control module; a radio control unit for adjusting a gain of the digital power control module and for adjusting a gain of the analog power control module; and a power suppression unit coupled to the radio control unit for providing power control commands to the digital power control module.

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According to a further aspect of the invention, a method for power control in a radio communication system is provided. The method includes determining a power requirement for an output signal; comparing the power requirement with a threshold power value; variably modifying the output signal in a digital domain
5 and in an analog domain; and modifying the power requirement when the power requirement is greater than the threshold power value such that the output signal has a modified power requirement which is less than or equal to the threshold power value.

According to a further aspect of the invention, a method of power control
10 in a radio communication system is provided. The method includes modulating a plurality of baseband data streams; adjusting a power level of each of the plurality of modulated baseband data streams; upconverting, to a respective carrier frequency, each of the plurality of power modulated baseband data streams to form a plurality of individual carriers; combining the plurality of individual
15 carriers into a single data stream with a power level requirement which is a function of the power levels associated with the plurality of individual carriers; modifying the power level requirement of the single data stream; converting the single data stream into an analog waveform; and variably adjusting the power level of the analog waveform.

20 It is desirable to limit P_{MCPA} by reducing the margin between the average transmit power over several time slots and P_{MCPA} in order to, for example, reduce the cost of the MCPA. The power control method of the present invention allows for a lower P_{MCPA} , and, therefore, a lower cost MCPA to be employed by a base station. In addition, the present invention does not need to sacrifice quality in
25 order to reduce costs since it is able to maintain sufficient average power.

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BRIEF DESCRIPTION OF THE DRAWINGS

The above objects and features of the present invention will be more apparent from the following description of the preferred embodiments with reference to the accompanying drawings, wherein:

- 5 FIG. 1 illustrates a conventional base station employing a MCPA;
 FIG. 2 illustrates an exemplary time chart which may be associated with a base station;
 FIG. 3 illustrates an exemplary cellular mobile radio system;
 FIG. 4 illustrates an exemplary probability model utilized by the present
10 invention;
 FIGS. 5A-5B illustrate the power control system according to exemplary embodiments of the present invention; and
 FIGS. 6A-6F illustrate power control methods according to exemplary embodiments of the present invention.

15 DETAILED DESCRIPTION

 FIG. 3 illustrates an exemplary cellular mobile radio system. In FIG. 3, the cellular mobile radio system is illustrated as comprising ten cells. One skilled in the art will appreciate that a typical cellular mobile radio system commonly includes more than ten cells; however, for the sake of brevity, the present
20 invention is explained using the simplified representation illustrated in FIG. 3.

 Each cell C1-C10 has a corresponding base station B1-B10. FIG. 3 illustrates the base stations as situated in the vicinity of the cell center and having omni-directional antennas. The cells C1-C10 are, therefore, schematically represented as hexagons. The base stations of adjacent cells may, however, be co-
25 located in the vicinity of cell borders and have directional antennas as is well known to those skilled in the art. The base stations, according to exemplary embodiments of the present invention, are equipped with MCPAs and operate using time division multiple access (TDMA) techniques. In an additional

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exemplary embodiment of the present invention, a single MCPA is shared by multiple co-located base stations.

FIG. 3 also illustrates nine mobile stations M1-M9, moveable within a cell and among a plurality of cells. In a typical cellular radio system there would normally be more than nine cellular mobile stations. In fact, there are typically many times the number of mobile stations as there are base stations. However, for the purpose of explaining the invention, the reduced number of mobile stations is sufficient.

Also illustrated in FIG. 3 is an exemplary number of base station controllers BSC1-BSC3 and a mobile switching center MSC. The base station controllers BSC1-BSC3 control the operation of the base stations. The mobile switching center MSC is connected to each of the base station controllers BSC1-BSC3 and to a fixed public switching telephone network or similar fixed network (not shown). While only one mobile switching center is illustrated in FIG. 3, one skilled in the art will appreciate that additional mobile switching centers may be employed. In addition, one skilled in the art will also appreciate that exemplary cellular mobile radio systems can also operate without the use of base station controllers.

As mentioned above, the present invention is most applicable to a TDMA environment where the power of a system is limited. An exemplary base station in which the method of the present invention may be implemented was set forth above with respect to FIG. 1. Such a base station, operating in a TDMA environment, is commonly associated with a plurality of frequency carriers. Each frequency carrier is divided into a plurality of time slots to which mobile stations are assigned. Returning to FIG. 2, for example, the frequencies associated with an exemplary base station are depicted. Each of the seven frequencies is divided, for explanatory purposes, into eight time slots. Mobile users would be assigned to one or more time slots for transmission purposes.

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When the MCPA has been calibrated to a particular maximum tolerable output power, P_{MCPA} , exceeding that maximum limit results in a loss in linearity. A loss in linearity can cause an increase in spurious emissions from, for example, intermodulation products. P_{MCPA} can be limited by any sub-component within the transmitter chain. That is, the maximum tolerable power can be reduced by, for example, limitations of amplifiers, filters, digital-to-analog converters, and/or upconverters. Therefore, it is desirable to limit P_{MCPA} by reducing the margin between the average transmit power over several time slots and P_{MCPA} in order to, for example, reduce the cost of the MCPA. The power control method of the present invention allows for a lower P_{MCPA} , and, therefore, a lower cost MCPA to be employed by a base station. In addition, the present invention does not need to sacrifice quality in order to reduce costs since it is able to maintain sufficient average power.

Alternatively, the MCPA of the present invention can accommodate more users than conventional MCPAs by an efficient use of the MCPA's resources.

The probability that all of the users on the same time slot of a base station are active is given by:

$$P_r(AllActive) = \prod_{n=1}^N x_n \approx x^N \quad (2)$$

where N is the number of users per time slot and x is the speech activity per user. Speech activity is typically measured by detecting talk spurts and silence periods. Assuming that all users have the same speech activity, then $P_r(AllActive) = x^N$. For example, with eight users each having a 50% speech activity, the probability of all users being active is as low as 1/256 or 0.4%.

FIG. 4 illustrates a model of the probability of talk spurts and silence periods for exemplary users of the present invention. The probability for a user to be in a talk spurt is given by:

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$$P_r(Talk) = \frac{p}{q+p} \quad (3)$$

where p is the probability of the user changing states from a silence period to a talk spurt and q is the probability of user changing from a talk spurt to a silence period. The probability for a user being in a silent period is given by:

$$P_r(Silence) = \frac{q}{q+p} \quad (4)$$

Using equations (3) and (4) and taking into account that calls made by each user are independent of each other, the probability that at least n out of y users are active is given by:

$$P_r(X \geq n) = \sum_{x=n}^y \binom{y}{x} \left(\frac{p}{q+p} \right)^x \left(\frac{q}{q+p} \right)^{y-x} \quad (5)$$

For example, the probability of n out of y users being active when n approaches y is extremely low. Therefore, even if the maximum transmit power of each user is set such that in the worst case (i.e., all users are active at the same time) P_{MCPA} is exceeded, the actual probability of exceeding P_{MCPA} will be extremely low. However, even though the probability is extremely low, there will be short periods where the desired output power will exceed P_{MCPA} .

These short periods in which the desired output power will exceed P_{MCPA} are handled by an exemplary embodiment of the present invention as illustrated in FIG. 5A. FIG. 5A illustrates a MCPA 500 according to an exemplary embodiment of the present invention. Similar to multicarrier transmitter 100 of FIG. 1, each of a number of baseband frequency data signals $BB_1 \dots BB_N$ are forwarded to a corresponding modulator $Mod_1 \dots Mod_N$. The modulated baseband signals $BB_1 \dots BB_N$ are then forwarded to the digital power control modules

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DPC₁...DPC_N. The Radio Control Unit 5650 individually sets the gain of each digital power control module DPC₁...DPC_N.

The Radio Control Unit 5650 receives power command signals 5653 from a Peak Power Suppression Unit 5652. The signals 5653 are power control
 5 commands informing the Radio Control Unit 5650 of the actual output level at the transmitter, $P_{k,out}$, for each carrier and time slot. The Peak Power Suppression Unit 5652 calculates, on a time slot basis, the desired P_{sum} by assuming the power control commands 5651 would not be modified by the Peak Power Suppression Unit 5652. The Peak Power Suppression Unit 5652 modifies the power control
 10 commands for one or several users, if the desired P_{sum} exceeds the maximum tolerable threshold level for the MCPA, P_{MCPA} , in order to keep P_{sum} less than or equal to P_{MCPA} . The actual desired power level signals 5651 are obtained from a conventional open or closed loop power control algorithm.

The signals generated by the digital power control modules DPC₁...DPC_N
 15 are forwarded to multipliers Multi₁...Multi_N where the signals are upconverted to a corresponding carrier frequency. The upconverted signals are then summed by adder 5610. The resultant compound signal is converted from digital to analog by DAC 5620. The analog signal is forwarded to an analog power control module 5660 which has a constant gain, G_{ana} . Although analog power control module
 20 5660 is illustrated as a single element in FIG. 5A, the functionality of the analog power control module 5660 may be distributed over the analog transmitter chain, i.e., amplifiers, filters and upconverters, in order to minimize the effect on the noise figure.

The Peak Power Supervision Unit 5652 and the Radio Control Unit 5650
 25 can be implemented using any known digital signal processing technology, including a general purpose computer, a collection of integrated circuits and/or application specific integrated circuitry (ASIC).

FIG. 5B illustrates a MCPA 500' according to another exemplary embodiment of the present invention. Similar to multicarrier transmitter 100 of

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FIG. 1, each of a number of N baseband frequency data signals $BB_1...BB_N$ are forwarded to a corresponding modulator $Mod_1...Mod_N$. The modulated baseband signals $BB_1...BB_N$ are then forwarded to the digital power control modules $DPC_1...DPC_N$. The Radio Control Unit 5650' individually sets the gain of each digital power control module $DPC_1...DPC_N$, such that the total in-phase power sum in to the DAC 5620 equals the full scale range of the DAC, $P_{FS,DAC}$.

The Radio Control Unit 5650' receives power command signals 5653 from a Peak Power Suppression Unit 5652. The signals 5653 are power control commands informing the Radio Control Unit 5650' of the actual output level at the transmitter, $P_{k,out}$, for each carrier and time slot. The Peak Power Suppression Unit 5652 calculates, on a time slot basis, the desired P_{sum} by assuming the power control commands 5651 would not be modified by the Peak Power Suppression Unit 5652. The Peak Power Suppression Unit 5652 modifies the power control commands for one or several users, if the desired P_{sum} exceeds the maximum tolerable threshold level for the MCPA, P_{MCPA} , in order to keep P_{sum} less than or equal to P_{MCPA} . The actual desired power level signals 5651 are obtained from a conventional open or closed loop power control algorithm.

The signals generated by the digital power control modules $DPC_1...DPC_N$ are forwarded to multipliers $Multi_1...Multi_N$ where the signals are upconverted to a corresponding carrier frequency. The upconverted signals are then summed by adder 5610. The resultant compound signal is converted from digital to analog by DAC 5620. The analog signal is forwarded to an analog power control module 5660' which is controlled by an analog power control table stored in the Radio Control Unit 5650'. Although analog power control module 5660' is illustrated as a single element in FIG. 5B, the functionality of the analog power control module 5660' may be distributed over the analog transmitter chain, i.e., amplifiers, filters and upconverters, in order to minimize the effect on the noise figure.

In order to correctly adjust the absolute power level at the transmitter, $P_{k,out}$, for the carrier signals, the power level of the carrier signals are all adjusted

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by the same amount by the analog power control module 5660'. In order to adjust the gain of each digital power control module and the gain of the analog power control module 5660', the Radio Control Unit 5650' obtains, from the power control command signals 5653, the required power level of each carrier at the output of the transmitter $P_{k,out}$, i.e., the amount of transmit power required to provide a corresponding mobile station with an acceptable signal. Since the full scale range of the DAC 5620 is determined at the time of installation of the DAC 5620 in the multicarrier transmitter, the full scale range $P_{FS,DAC}$ of the DAC 5620 and the corresponding code can be programmed into the Radio Control Unit 5650'. The Radio Control Unit 5650' then sets the relative power for all carriers in such a way that the in-phase power sum in the digital domain supplied to the DAC 5620 equals the DAC's full scale range. Thus, in order to obtain the correct actual output power, $P_{k,out}$, at the transmitter for each carrier, Radio Control Unit 5650' applies an adjustable analog power gain, G_{ana} , for the analog power amplifier 5660'. Thus, all carriers will always be amplified by the same amount, G_{ana} , but G_{ana} can change from time slot to time slot.

FIG. 6A illustrates a flow diagram of an exemplary embodiment of the present invention. The Peak Power Supervision Unit 5650 (see FIGS. 5A-5B) selects the present time slot i in block 670. The desired in-phase power sum for time slot i , P_{sum} , is calculated in block 672. In block 674, the value of P_{sum} is compared to P_{MCPA} . If $P_{sum} \leq P_{MCPA}$ then the individual power level commands are allowed to pass to the Radio Control Unit 5650, 5650' (see FIGS. 5A-5B) without manipulation or adjustment (block 680). In block 678, i is incremented by one ($i = (i+1) \bmod T$), where T represents the number of time slots for the system (e.g., $T = 8$ for a GSM system) and the above procedure is performed on the new time slot i .

If, however, $P_{sum} > P_{MCPA}$ then the present invention proceeds to block 676. In block 676, P_{sum} is lowered and the present invention returns to block 672 and repeats blocks 674-676 until P_{sum} is less than or equal to P_{MCPA} .

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FIG. 6B illustrates an exemplary embodiment for block 676 of FIG. 6A. In block 676B₁, a user k is selected out of the total number of users on the time slot i ($[1, \dots, M]$). In block 676B₂, the power of a single speech burst for user k is set to zero. For the following speech burst, the power of a user $k+1$ is set to zero. This continues such that each user has their power set to zero for example, 1 (or more) out of 8 bursts per speech frame. Conventional error correction techniques, such as speech coding and interleaving, are used to mitigate the effects caused by the loss of a single speech burst so as to not compromise overall speech quality. The exemplary method of FIG. 6B effectively spreads quality reductions over multiple users so as to reduce the impact as much as possible on the perceived quality by any given user.

FIG. 6C illustrates another exemplary embodiment for block 676 of FIG. 6A. In block 676C₁, a user k is selected out of the total number of users on the time slot i ($[1, \dots, M]$). Then, in block 676C₂, the power of user k , P_k is reduced. The present invention returns to block 672 of FIG. 6A and repeats blocks 674-676 selecting different values of k until P_{sum} is less than or equal to P_{MCPA} .

FIG. 6D illustrates another exemplary embodiment for block 676 of FIG. 6A. In block 676D₁, all users k are selected out of the total number of users on the time slot i ($[1, \dots, M]$). Then in block 676D₂, the power of each user is either reduced equally or dependent to the power of each user. That is, the power of each user can be reduced by the same percentage (e.g., 20% dB or 20% W) or the same value (e.g., 1 dB or 1 W). Alternatively, the power of each user can be reduced dependent on their current power. For example, users which have higher powers are usually further away from a base station than users with lower powers. A reduction in power for a far-user could reduce signal quality more for the far-user than the same power reduction made for a near-user. Therefore, an exemplary embodiment of the present invention reduces the power dependent on the present power such that the power of near-users are reduced more than the

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power of far-users, e.g., inversely proportional to their present power. In a further exemplary embodiment of the present invention, the power of far-users are reduced more than the power of near-users since signals to the far-users would most probably be transmitted at a higher power than signals transmitted to the near-users.

FIG. 6E illustrates another exemplary embodiment for block 676 of FIG. 6A. In block 676E₁, the user k which has the greatest power is selected. Then in block 676E₂, the power of user k is reduced.

FIG. 6F illustrates another exemplary embodiment for block 676 of FIG. 6A. In block 676F₁, the user k which has the greatest perceived signal quality is selected. Alternatively, the user k which has the weakest perceived signal quality may be selected. Then in block 676F₂, the power of user k is reduced. The perceived signal quality can be determined, for example, by calculating a signal-to-noise ratio or bit-error rate.

The foregoing has described the principles, preferred embodiments and modes of operation of the present invention. However, the invention should not be construed as being limited to the particular embodiments discussed above. For example, while the invention has been described in the context of a TDMA system, it is understood the present invention is usable in the code division multiple access (CDMA) and frequency division multiple access (FDMA) systems as well. Thus, the above-described embodiments should be regarded as illustrative rather than restrictive, and it should be appreciated that variations may be made in those embodiments by workers skilled in the art without departing from the scope of the present invention as defined by the following claims.

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WE CLAIM:

1. A method for power control in a radio communication system comprising:
determining a power requirement for an output signal;
comparing said power requirement with a threshold power value; and
5 modifying said output signal when said power requirement is greater than
said threshold power value such that the modified output signal has a modified
power requirement which is less than or equal to said threshold power value.
2. The method of claim 1, wherein said power requirement is a desired power
requirement.
- 10 3. The method of claim 1 further comprising:
providing as an input to a multi-carrier power amplifier said modified
output signal.
4. The method of claim 1, wherein said threshold power value is a threshold
power value for a multi-carrier power amplifier.
- 15 5. The method of claim 1, wherein said threshold power value is a threshold
power value for a component in a transmitter chain of said radio communication
system.
6. The method of claim 1, wherein said radio communication system is a
TDMA system, said system having multiple time slots.
- 20 7. The method of claim 1, wherein said comparing step further comprises
comparing a power requirement for at least one time slot of said output signal to
said threshold power value; and

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wherein said modifying step further comprises modifying said at least one time slot of said output signal when said power requirement of said at least one time slot of said output signal is greater than said threshold power value such that the modified at least one time slot of said output signal has a modified power
5 requirement which is less than or equal to said threshold power value.

8. The method of claim 6, wherein said radio communication system includes a plurality of mobile terminals, said plurality of mobile terminals receiving over a plurality of radio carriers, said method further comprising:
transmitting information to at least one of said plurality of mobile terminals
10 on successive time slots.

9. The method of claim 6, wherein said radio communication system includes a plurality of mobile terminals, said plurality of mobile terminals receiving over a plurality of radio carriers, each at a respective power level, said method further comprising:
15 transmitting information to at least one of said plurality of mobile terminals by interleaving said information on said time slots.

10. The method of claim 9, wherein said output signal is modified by reducing the respective power level for at least one time slot for said at least one of said plurality of mobile terminals.

20 11. The method of claim 6, wherein said radio communication system includes a plurality of mobile terminals, said plurality of mobile terminals receiving over a plurality of radio carriers, each at a respective power level, wherein said output signal is modified by reducing the respective power levels for all of said plurality of mobile terminals.

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12. The method of claim 6, wherein said radio communication system includes a plurality of mobile terminals, said plurality of mobile terminals receiving over a plurality of radio carriers, each at a respective power level, wherein said output signal is modified by reducing the respective power levels for all of said plurality of mobile terminals dependent on their respective power levels.

13. The method of claim 12, wherein said output signal is modified by reducing the respective power levels for said plurality of mobile terminals inversely proportional to their respective power levels.

14. The method of claim 6, wherein said radio communication system includes a plurality of mobile terminals, said plurality of mobile terminals receiving over a plurality of radio carriers, each at a respective power level, wherein said output signal is modified by reducing the respective power level for at least one of said plurality of mobile terminals.

15. The method of claim 6, wherein said radio communication system includes a plurality of mobile terminals, said plurality of mobile terminals receiving over a plurality of radio carriers, each at a respective power level, wherein said output signal is modified by reducing the respective power level for one of said plurality of mobile terminals which has a greater respective power level than any of the other plurality of mobile terminals.

16. The method of claim 6, wherein said radio communication system includes a plurality of mobile terminals, said plurality of mobile terminals receiving over a plurality of radio carriers, each at a respective power level, wherein said output signal is modified by reducing the respective power level for one of said plurality of mobile terminals based on a respective signal quality value.

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17. A method for power control in a radio communication system, said radio communication system including a power amplifier, a plurality of mobile terminals, said plurality of mobile terminals receiving over a plurality of radio carriers, each at a respective power level, said method comprising:

5 allocating said respective power level for at least one of said plurality of mobile terminals such that an instantaneous power of an output signal for said plurality of mobile terminals will exceed a threshold power value of said power amplifier;

determining a power requirement for said output signal;

10 comparing said power requirement for said output signal with said threshold power value; and

modifying said output signal when said power requirement is greater than said threshold power value such that the modified output signal has a modified power requirement which is less than or equal to said threshold power value.

15 18. A power control system for a radio communication system comprising:
a power supervision unit which determines a power requirement for an output signal and compares said power requirement with a threshold power value;
and

20 a signal control unit which modifies said output signal when said power requirement is greater than said threshold power value such that the modified output signal has a modified power requirement which is less than or equal to said threshold power value.

19. The system of claim 18, wherein said modified output signal is provided as an input to a multi-carrier power amplifier.

25 20. The system of claim 18, wherein said threshold power value is a threshold power value for a multi-carrier power amplifier.

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21. The system of claim 18, wherein said threshold power value is a threshold power value for a component in a transmitter chain of said radio communication system.

5 22. The system of claim 18, wherein said radio communication system is a TDMA system, said system having multiple time slots.

23. The system of claim 18, wherein said power supervision unit compares a power requirement for a at least one time slot of said output signal to said threshold power value; and

10 wherein said signal control unit modifies said at least one time slot of said output signal when said power requirement of said at least one time slot of said output signal is greater than said threshold power value such that the modified at least one time slot of said output signal has a modified power requirement which is less than or equal to said threshold power value.

15 24. The system of claim 22, wherein said radio communication system includes a plurality of mobile terminals, said plurality of mobile terminals receiving over a plurality of radio carriers, and wherein said radio communication system transmits information to at least one of said plurality of mobile terminals on consecutive time slots.

20 25. The system of claim 22, wherein said radio communication system includes a plurality of mobile terminals, said plurality of mobile terminals receiving over a plurality of radio carriers, each at a respective power level, and wherein said radio communication system transmits information to at least one of said plurality of mobile terminals by interleaving said information on said time slots.

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26. The system of claim 24, wherein said output signal is modified by reducing the respective power level of at least one time slot of said at least one of said plurality of mobile terminals.
- 5 27. The system of claim 22, wherein said radio communication system includes a plurality of mobile terminals, said plurality of mobile terminals receiving over a plurality of radio carriers, each at a respective power level, wherein said output signal is modified by reducing the respective power level for all of said plurality of mobile terminals.
- 10 28. The system of claim 22, wherein said radio communication system includes a plurality of mobile terminals, said plurality of mobile terminals receiving over a plurality of radio carriers, each at a respective power level, wherein said output signal is modified by reducing the respective power levels for all of said plurality of mobile terminals dependent on their respective power levels.
- 15 29. The system of claim 28, wherein said output signal is modified by reducing the respective power levels for said plurality of mobile terminals inversely proportional to their respective power levels.
- 20 30. The system of claim 22, wherein said radio communication system includes a plurality of mobile terminals, said plurality of mobile terminals receiving over a plurality of radio carriers, each at a respective power level, wherein said output signal is modified by reducing the respective power level for at least one of said plurality of mobile terminals.
31. The system of claim 22, wherein said radio communication system includes a plurality of mobile terminals, said plurality of mobile terminals receiving over a plurality of radio carriers, each at a respective power level, wherein said output

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signal is modified by reducing the respective power level for one of said plurality of mobile terminals which has a greater respective power level than any of the other plurality of mobile terminals.

32. The system of claim 22, wherein said radio communication system includes
5 a plurality of mobile terminals, said plurality of mobile terminals receiving over a plurality of radio carriers, each at a respective power level, wherein said output signal is modified by reducing the respective power level for one of said plurality of mobile terminals based on a respective signal quality value.

33. A power control system for a radio communication system comprising:
10 means for allocating a respective power level for at least one of a plurality of mobile terminals such that an instantaneous power of an output signal for said plurality of mobile terminals will exceed a threshold power value of a power amplifier;

means for determining a power requirement for said output signal;
15 means for comparing said power requirement for said output signal with said threshold power value; and

means for modifying said output signal when said power requirement is greater than said threshold power value such that the modified output signal has a modified power requirement which is less than or equal to said threshold power
20 value.

34. A power control system for a radio communication system comprising:
a power supervision unit which determines a power requirement for an output signal and compares said power requirement with a threshold power value;
a signal control unit which variably modifies said output signal in a digital
25 domain and in an analog domain; and

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a power suppression unit coupled to said signal control unit which modifies said power requirement when said power requirement is greater than said threshold power value such that said output signal has a modified power requirement which is less than or equal to said threshold power value.

- 5 35. A power control system for a radio communication system, comprising:
 a digital power control module;
 a digital-to-analog converter;
 an analog power control module;
 a radio control unit for adjusting a gain of said digital power control
10 module and for adjusting a gain of said analog power control module; and
 a power suppression unit coupled to said radio control unit for providing
 power control commands to said digital power control module.

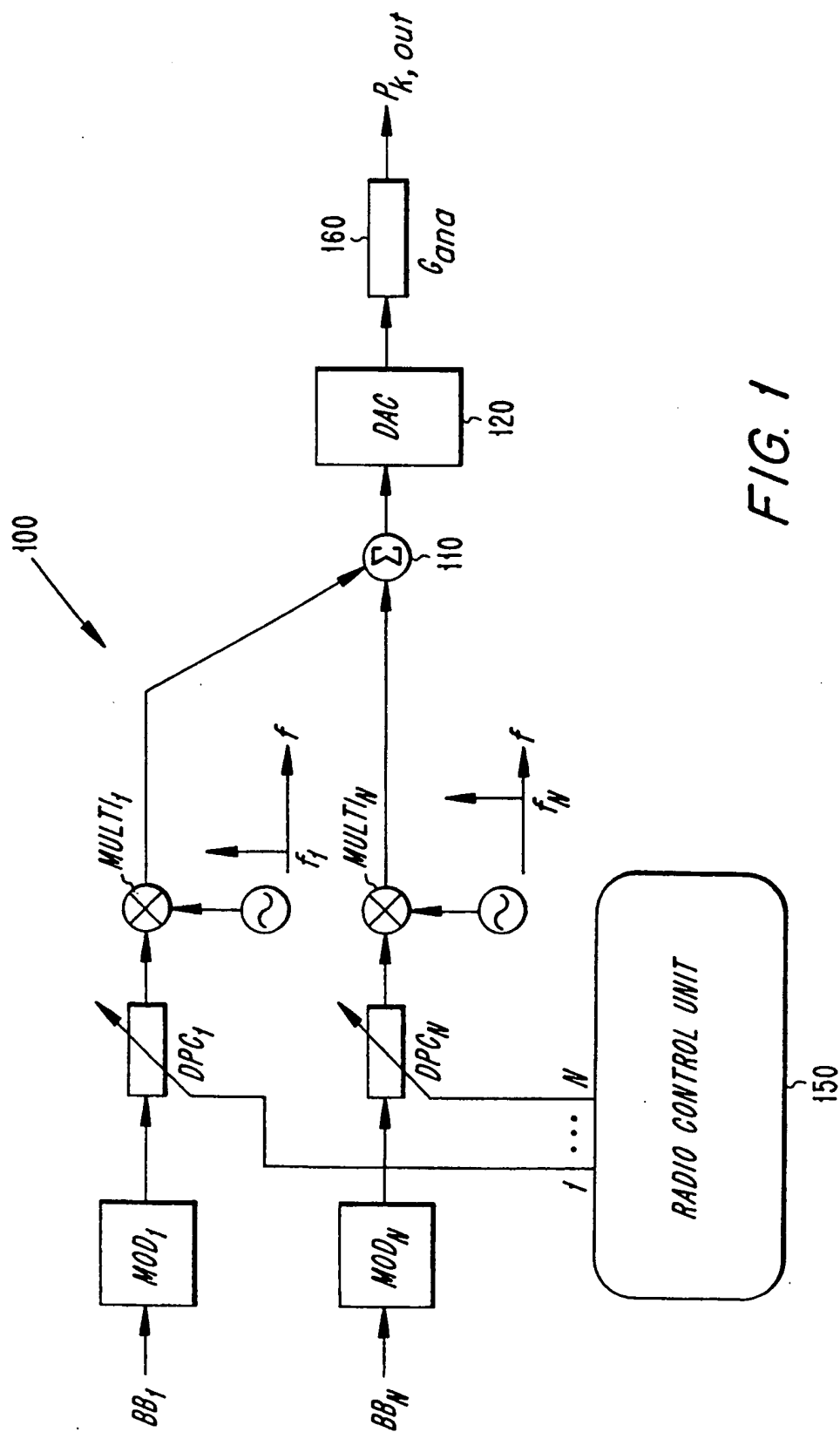
36. The system of claim 35, wherein said power control commands indicate a
 reduced power level requirement when an output signal has a desired power
15 requirement greater than a threshold value.

37. A method for power control in a radio communication system comprising:
 determining a power requirement for an output signal;
 comparing said power requirement with a threshold power value; and
 variably modifying said output signal in a digital domain and in an analog
20 domain;
 modifying said power requirement when said power requirement is greater
 than said threshold power value such that said output signal has a modified power
 requirement which is less than or equal to said threshold power value.

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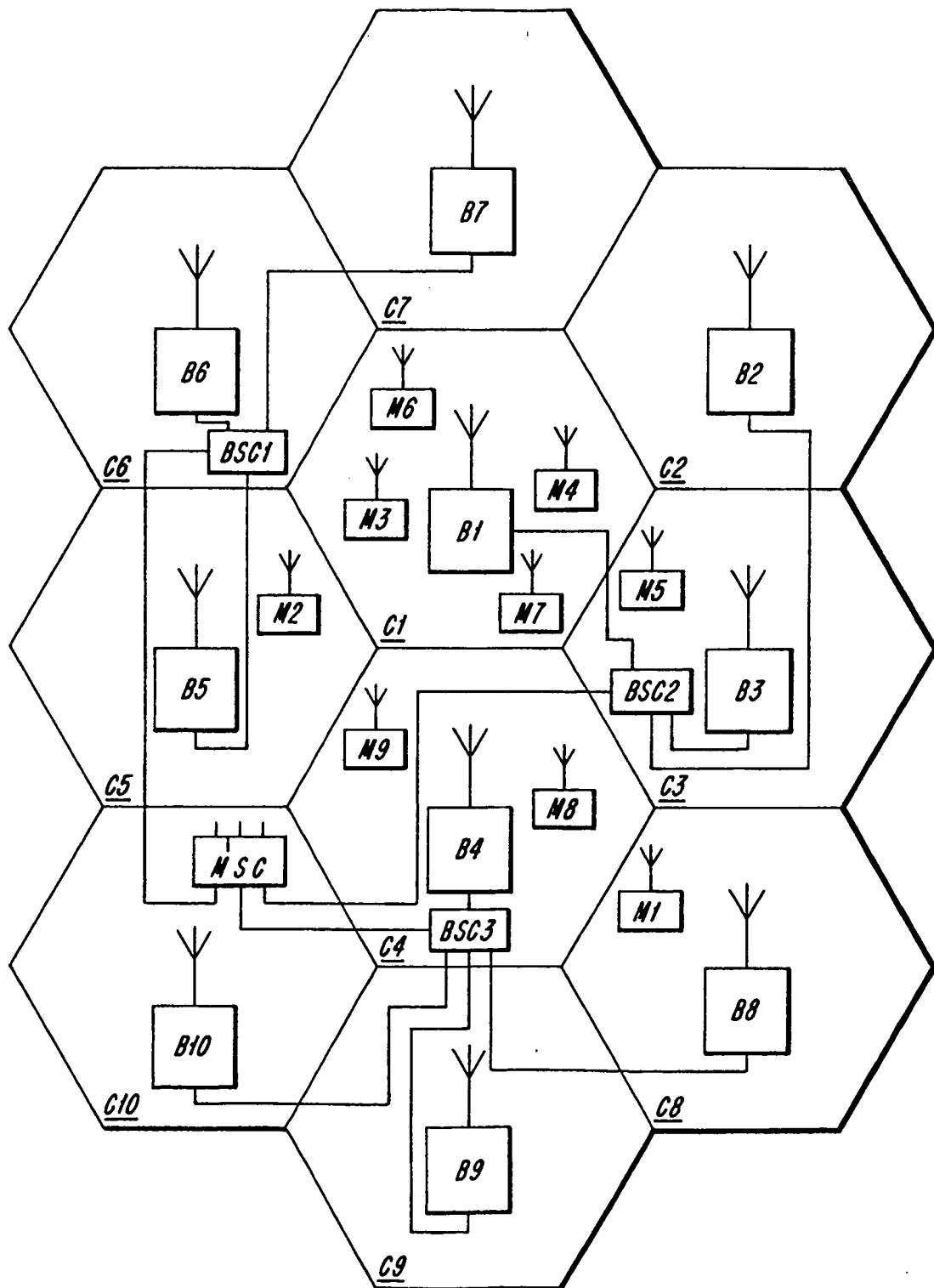
38. A method for power control in a radio communication system comprising:
modulating a plurality of baseband data streams;
adjusting a power level of each of said plurality of modulated baseband
data streams;
- 5 upconverting, to a respective carrier frequency, each of said plurality of
power modulated baseband data streams to form a plurality of individual carriers;
combining said plurality of individual carriers into a single data stream
with a power level requirement which is a function of the power levels associated
with said plurality of individual carriers;
- 10 modifying said power level requirement of said single data stream;
converting said single data stream into an analog waveform; and
variably adjusting the power level of said analog waveform.
39. The method of claim 38, wherein said power level requirement is modified
when said power level requirement of said single data stream will exceed a
15 threshold power value such that the modified power requirement is less than or
equal to said threshold power value.

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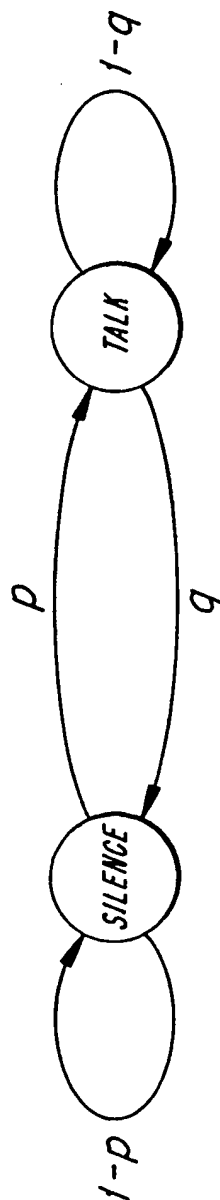
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FIG. 3



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FIG. 4



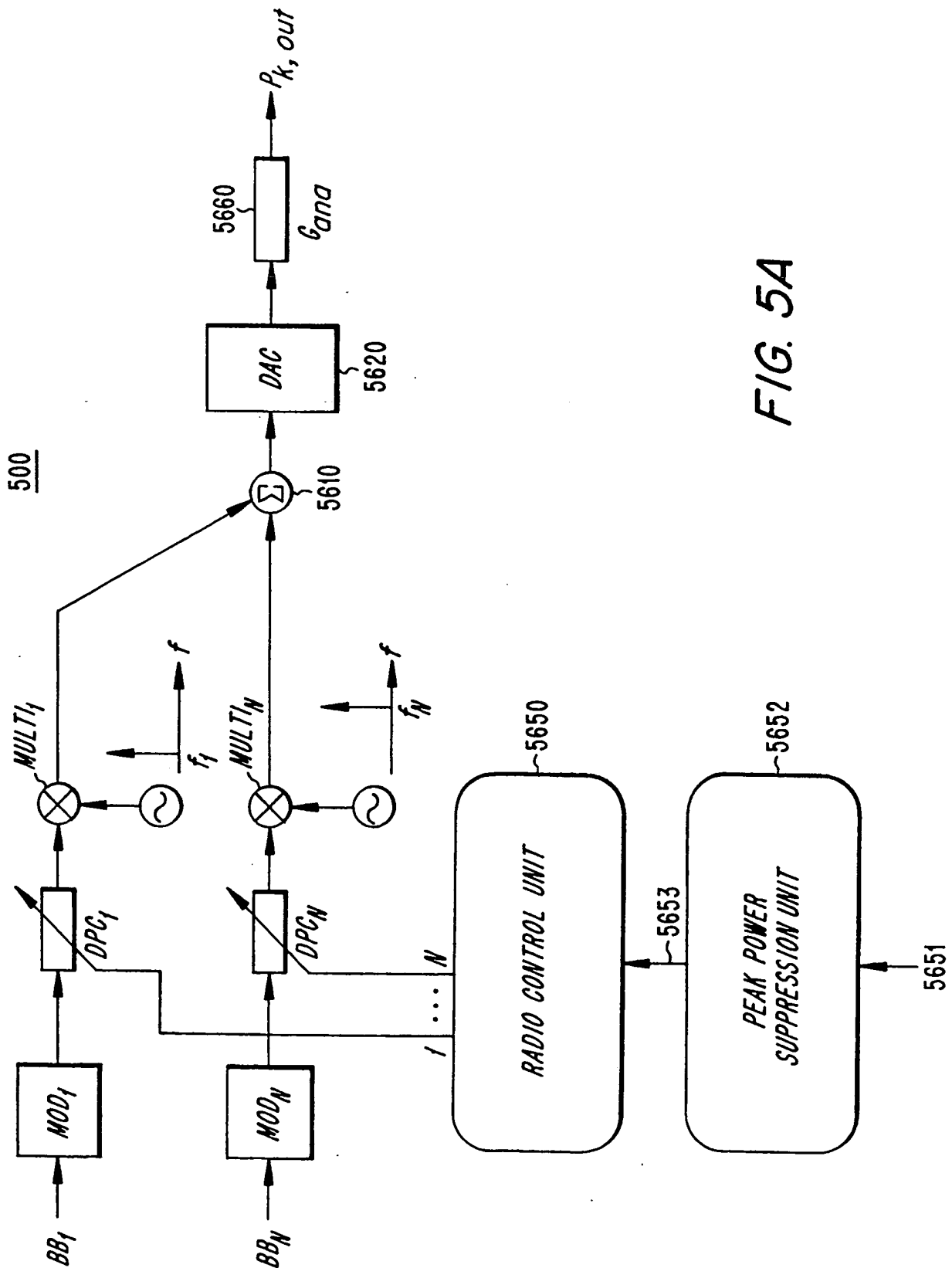


FIG. 5A

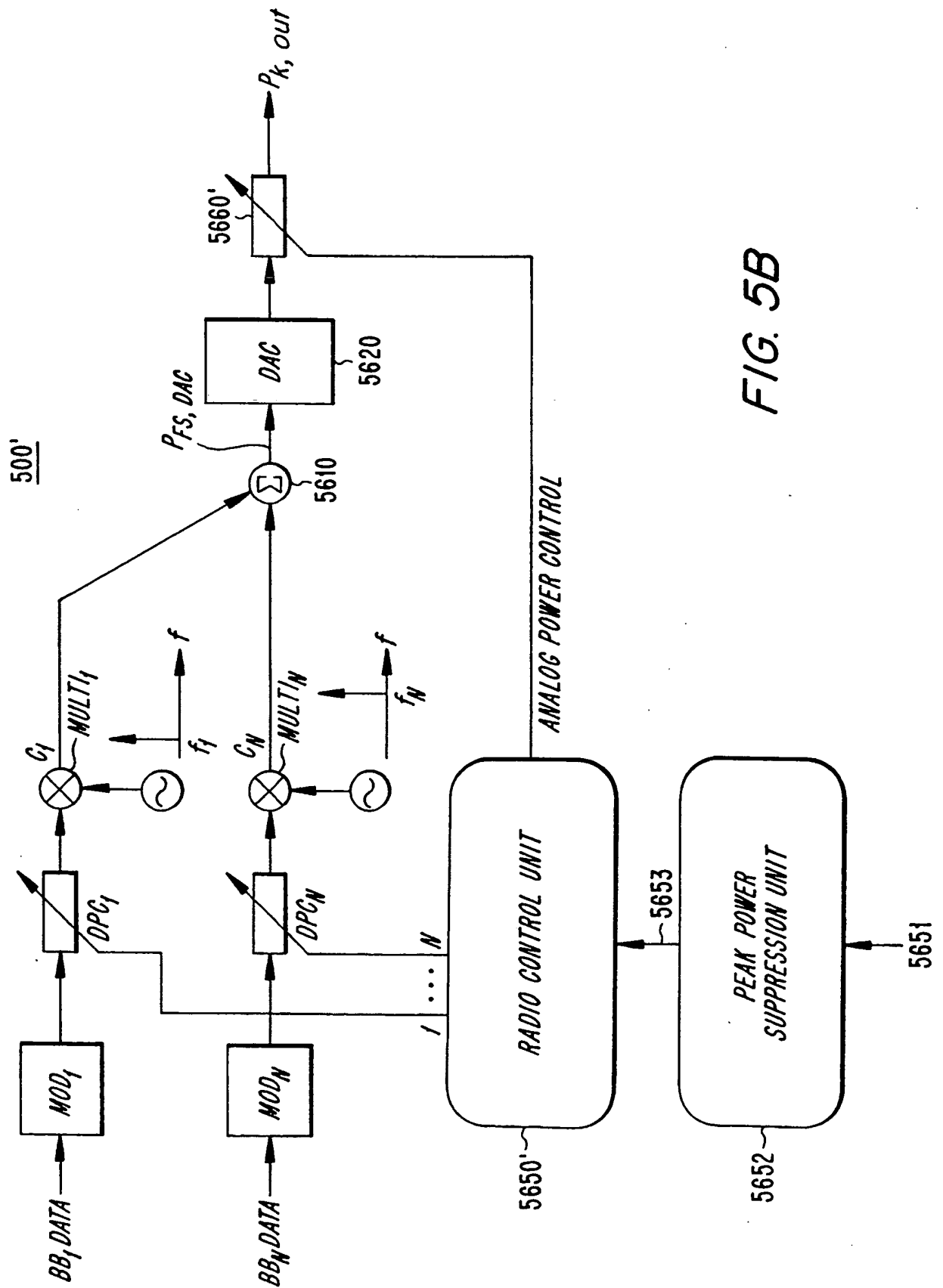


FIG. 5B

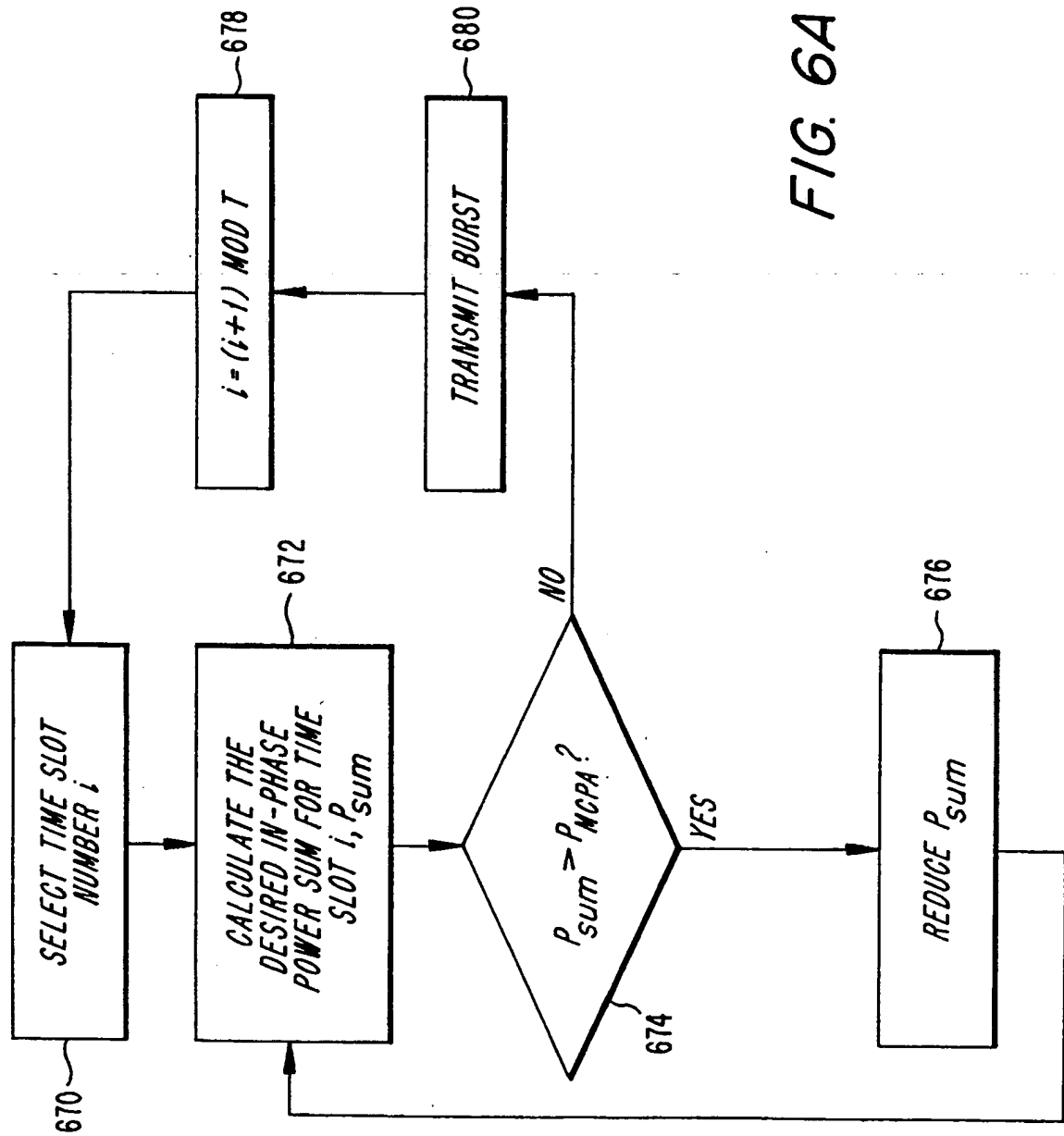


FIG. 6A

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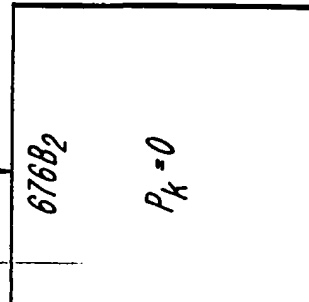
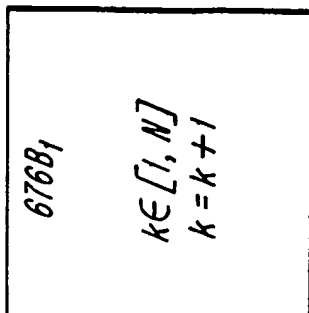


FIG. 6B

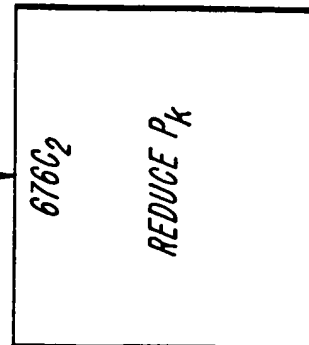
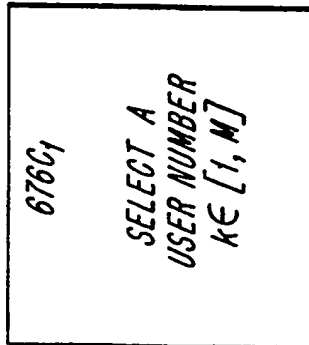


FIG. 6C

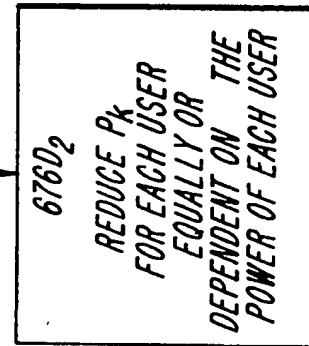
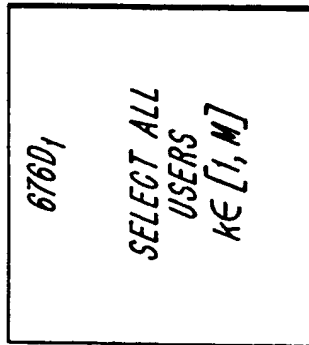


FIG. 6D

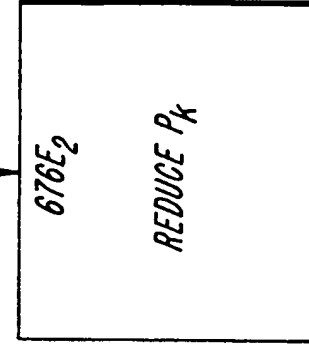
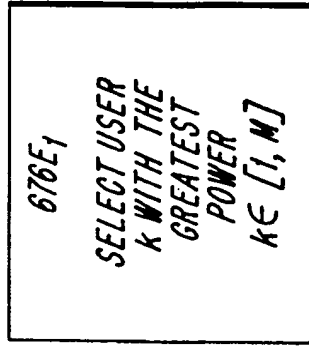


FIG. 6E

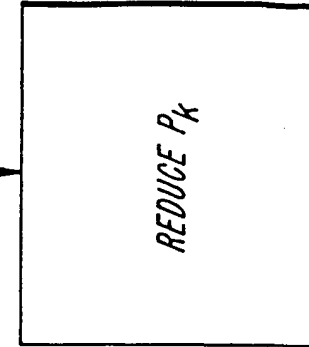
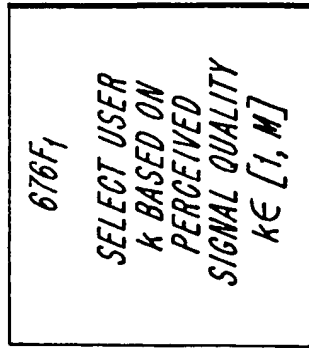


FIG. 6F

INTERNATIONAL SEARCH REPORT

International Application No

PCT/SE 00/01386

A. CLASSIFICATION OF SUBJECT MATTER

IPC 7 H04B7/005 H04L5/06 H04L27/26

According to International Patent Classification (IPC) or to both national classification and IPC

B. FIELDS SEARCHED

Minimum documentation searched (classification system followed by classification symbols)

IPC 7 H04B H04L

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Electronic data base consulted during the international search (name of data base and, where practical, search terms used)

EPO-Internal

C. DOCUMENTS CONSIDERED TO BE RELEVANT

Category *	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
X A	<p>EP 0 743 768 A (NIPPON TELEGRAPH & TELEPHONE) 20 November 1996 (1996-11-20)</p> <p>abstract column 3, line 13 - line 22 column 4, line 47 - column 5, line 9 column 5, line 19 - line 33 column 8, line 32 - column 9, line 2; figures 1,5,6 column 13, line 35 - column 14, line 14; figure 13 column 15, line 30 - column 17, line 5; figures 7-12,29</p> <p>-/--</p>	<p>1-11,14, 16-27, 30, 32-34, 37-39, 12,13, 15,28, 29,31, 35,36</p>



Further documents are listed in the continuation of box C.



Patent family members are listed in annex.

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Date of the actual completion of the international search

23 August 2000

Date of mailing of the international search report

05/09/2000

Name and mailing address of the ISA

European Patent Office, P.B. 5818 Patentlaan 2
NL - 2280 HV Rijswijk
Tel. (+31-70) 340-2040, Tx. 31 651 epo nl,
Fax: (+31-70) 340-3016

Authorized officer

Langinieux, F

C.(Continuation) DOCUMENTS CONSIDERED TO BE RELEVANT

Category *	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
	column 24, line 16 - last line; figures 15-28 ---	
X	WULICH D: "PEAK FACTOR IN ORTHOGONAL MULTICARRIER MODULATION WITH VARIABLE LEVELS" ELECTRONICS LETTERS, GB, IEE STEVENAGE, vol. 32, no. 20, 26 September 1996 (1996-09-26), pages 1859-1861, XP002042110 ISSN: 0013-5194	1,2, 6-11,14, 17,18, 22-27, 33,38,39
A	abstract page 1859, right-hand column, line 1 -page 1860, left-hand column, line 15 ---	3-5, 19-21, 34-37
X	US 5 384 547 A (LYNK JR CHARLES N ET AL) 24 January 1995 (1995-01-24) ---	1-6,11, 17-22, 27,33 34,37-39
A	abstract; figures 3,6 column 1, line 14 - line 52 ---	
X	US 5 894 498 A (SCHULER JOSEPH ET AL) 13 April 1999 (1999-04-13) ---	1-5, 18-21 17,33-39
A	abstract; figures 1-4 column 1, line 10 - line 43 column 2, line 27 -column 3, line 6 column 4, line 42 -column 5, line 6 column 11, line 13 - line 30 ---	
X	EP 0 719 001 A (ALCATEL BELL NV) 26 June 1996 (1996-06-26) ---	1,2,5, 18,21 3,4,17, 19,20, 33-39
A	abstract; figure 1 column 1, line 18 -column 2, line 6 -----	

Patent document cited in search report		Publication date	Patent family member(s)	Publication date
EP 0743768	A	20-11-1996	KR 195458 B	15-06-1999
			US 5790555 A	04-08-1998
			CA 2178815 A	06-06-1996
			CN 1138926 A	25-12-1996
			WO 9618249 A	13-06-1996
			JP 8274734 A	18-10-1996
			JP 2787619 B	20-08-1998
US 5384547	A	24-01-1995	NONE	
US 5894498	A	13-04-1999	CN 1248371 T	22-03-2000
			EP 0963643 A	15-12-1999
			WO 9838807 A	03-09-1998
EP 0719001	A	26-06-1996	AU 703482 B	25-03-1999
			AU 4041695 A	27-06-1996
			CA 2162326 A	23-06-1996
			IL 115799 A	27-12-1998
			US 5768318 A	16-06-1998